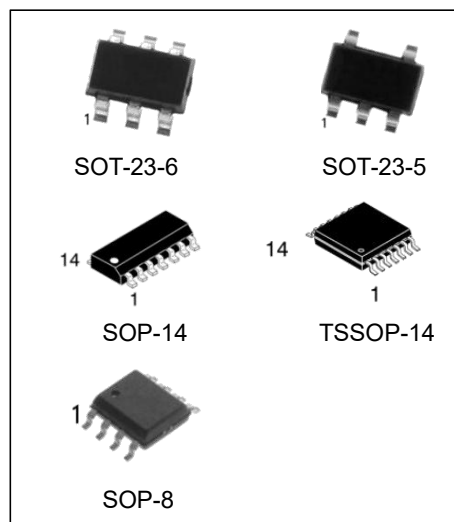


## 6.5MHz, 585μA, Rail-to-Rail I/O CMOS Operational Amplifier

### FEATURES

- LOW OFFSET: 5mV (max)
- LOW IB: 10pA (max)
- HIGH BANDWIDTH: 6.5MHz
- RAIL-TO-RAIL INPUT AND OUTPUT
- SINGLE SUPPLY: +2.3V to +5.5V
- SHUTDOWN: HXA373
- SPECIFIED UP TO +125°C
- MicroSIZE PACKAGES: SOT-23-5, SOT-23-6, and SOP-8



### Ordering Information

DEVICE	Package Type	MARKING	Packing	Packing Qty
HXA373AIDBVRG	SOT-23-6	A373/A75	REEL	3000pcs/reel
HXA373AIDRG	SOP-8	A373A	REEL	2500pcs/reel
HXA373AIDRG4	SOP-8	A373A	REEL	4000pcs/reel
HXA2373AIDGKRG	MSOP-10	A2373	REEL	3000pcs/reel
HXA374AIDBVRG	SOT-23-5	A374A/A76	REEL	3000pcs/reel
HXA374AIDRG	SOP-8	A374A	REEL	2500pcs/reel
HXA374AIDRG4	SOP-8	A374A	REEL	4000pcs/reel
HXA2374AIDBVRG	SOT-23-8	A2374A/ATP	REEL	3000pcs/reel
HXA2374AIDRG	SOP-8	A2374A	REEL	2500pcs/reel
HXA2374AIDRG4	SOP-8	A2374A	REEL	4000pcs/reel
HXA2374AIDGKRG	MSOP-8	A2374A	REEL	3000pcs/reel
HXA4374AIDRG	SOP-14	HXA4374A	REEL	2500pcs/reel
HXA4374AIDRG4	SOP-14	HXA4374A	REEL	4000pcs/reel
HXA4374AIPWRG	TSSOP-14	A4374A	REEL	2500pcs/reel

## **DESCRIPTION**

The HXA373A and HXA373 families of operational amplifiers are low power and low cost with excellent bandwidth (6.5MHz) and slew rate (5V/ $\mu$ s). The input range extends 200mV beyond the rails and the output range is within 25mV of the rails. Their speed/power ratio and small size make them ideal for portable and battery-powered applications.

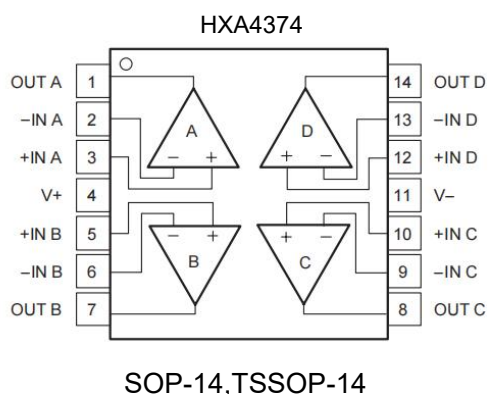
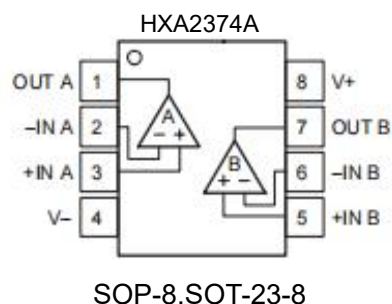
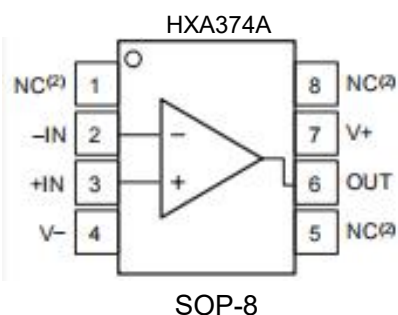
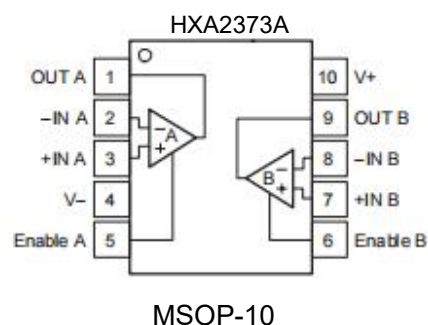
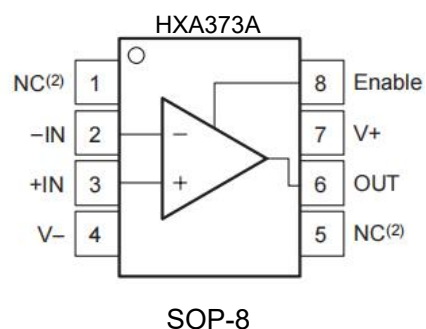
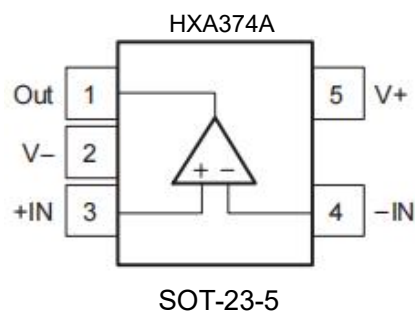
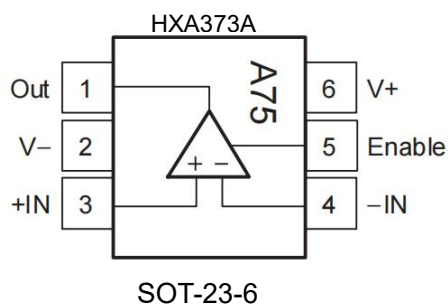
The HXA373A family includes a shutdown mode. Under logic control, the amplifiers can be switched from normal operation to a standby current that is less than 1 $\mu$ A.

The HXA373A and HXA373 families of operational amplifiers are specified for single or dual power supplies of +2.7V to +5.5V, with operation from +2.3V to +5.5V. All models are specified for  $-40^{\circ}\text{C}$  to  $+125^{\circ}\text{C}$ .

## **APPLICATIONS**

- Portable Equipment
- Battery-Powered Devices
- Active Filters
- Driving A/D Converters

## PIN ASSIGNMENT



(1) Pin 1 of the SOT23-6 is determined by orienting the package marking as shown.

(2) NC indicates no internal connection.

**ABSOLUTE MAXIMUM RATINGS(1)**

Condition	Min	Max	UNITS
Supply Voltage	-	+7.0	V
Signal Input Terminals,Voltage <sup>(2)</sup>	-0.5	+0.5	V
Signal Input Terminals,Current <sup>(2)</sup>	-10	+10	mA
Output Short-Circuit <sup>(3)</sup>	Continuous		
Operating Temperature	-55	+150	°C
Storage Temperature	-65	+150	°C
Junction Temperature	-	+150	°C
Lead Temperature (Soldering, 10 seconds)	-	245	°C

(1) Stresses above these ratings may cause permanent damage. Exposure to absolute maximum conditions for extended periods may degrade device reliability. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those specified is not implied.

(2) Input terminals are diode-clamped to the power-supply rails. Input signals that can swing more than 0.5V beyond the supply rails should be current-limited to 10mA or less.

(3) Short-circuit to ground, one amplifier per package.

**ELECTRICAL CHARACTERISTICS: VS = +2.7V to +5.5V**

**Boldface** limits apply over the specified temperature range, **TA = -40°C to +125°C**.

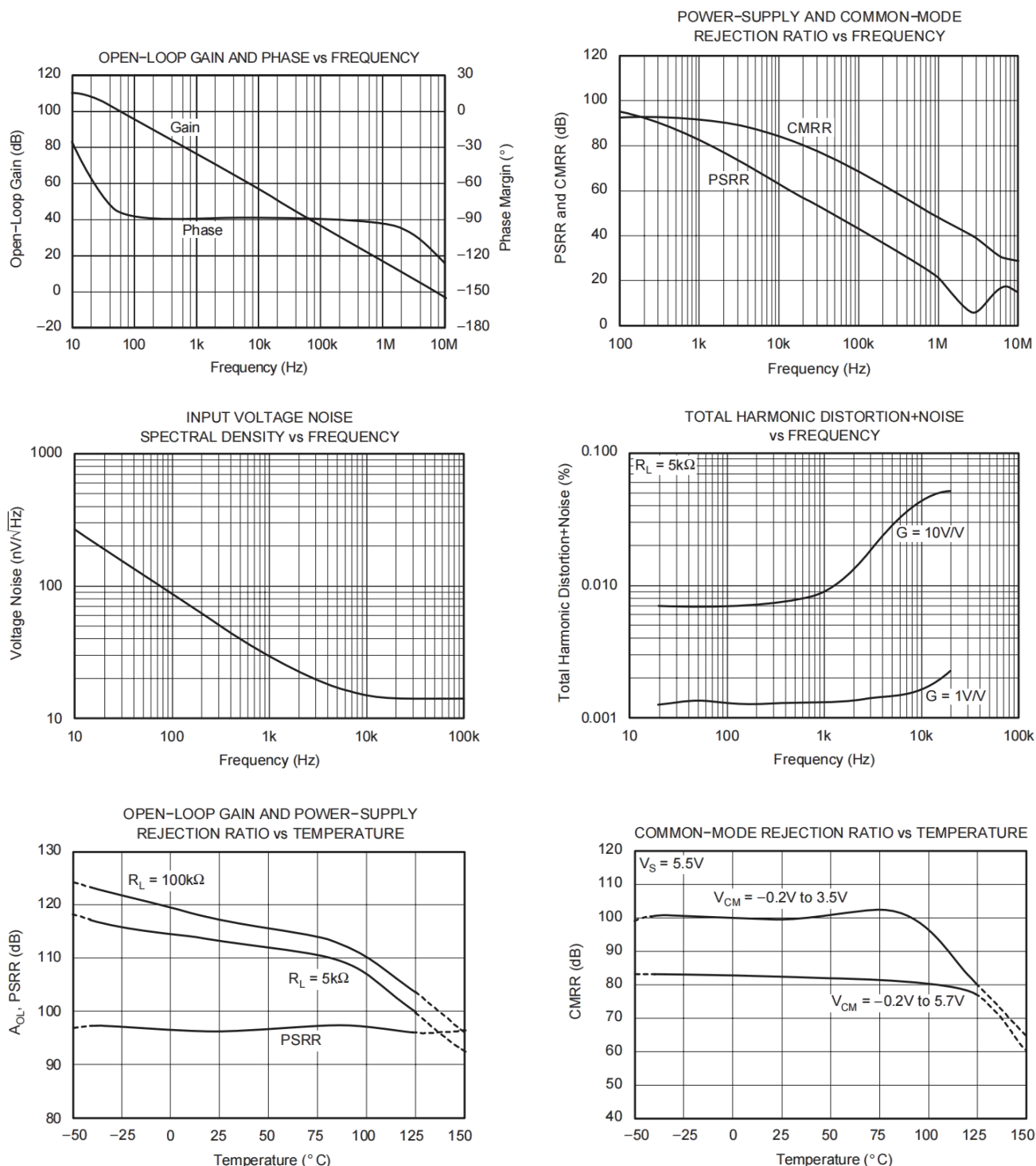
At TA = +25°C, RL = 10kΩ connected to VS/2, and VOUT = VS/2, unless otherwise noted.

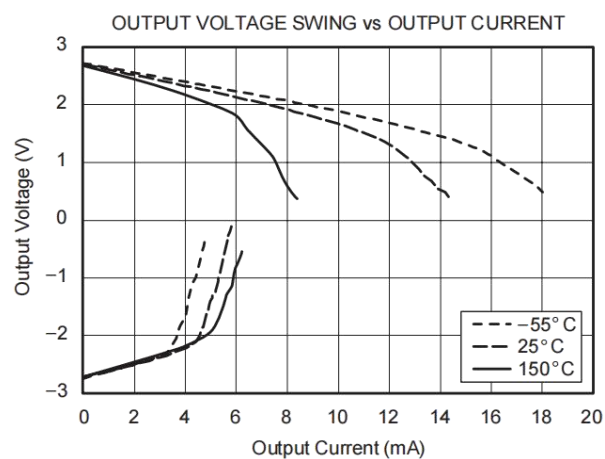
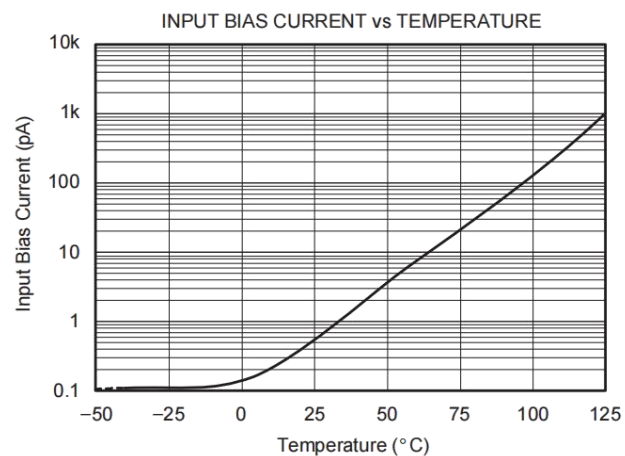
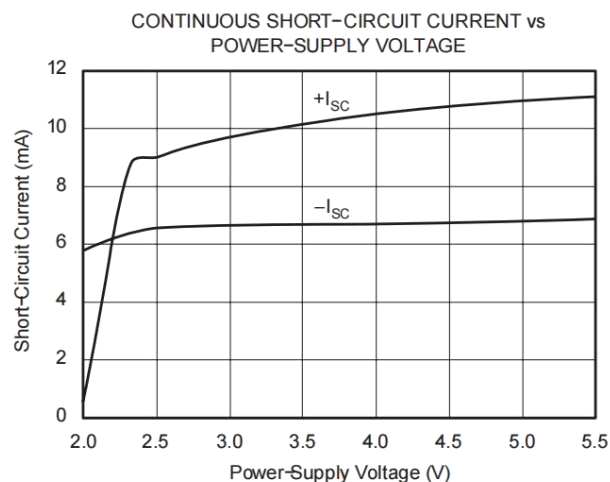
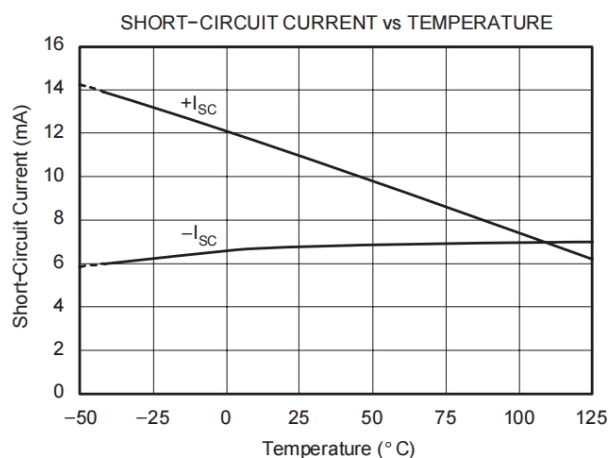
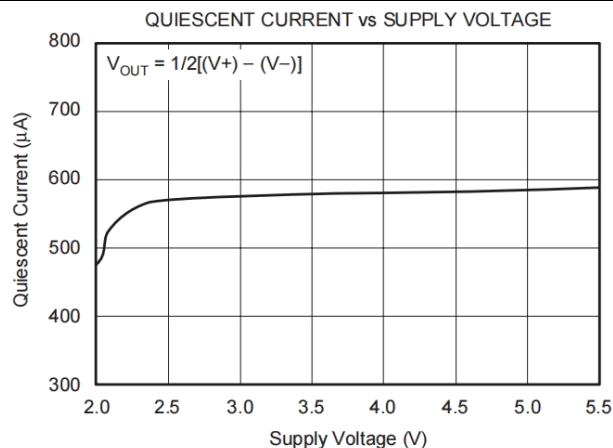
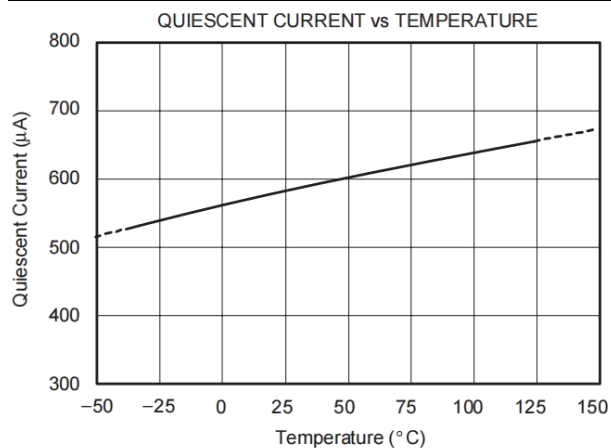
PARAMETER	CONDITIONS	HXA373A, HXA2373, HXA373, HXA2374, HXA4374			UNIT
		MIN	TYP	MAX	
<b>OFFSET VOLTAGE</b>					
Input Offset Voltage VOS	VS = 5V		1	5	mV
<b>over Temperature</b>				<b>6.5</b>	mV
<b>Drift</b> dVOS/dT			<b>3</b>		μV/°C
vs Power Supply PSRR	VS = 2.7V to 5.5V, VCM < (V+) - 2V		25	100	μV/V
<b>over Temperature</b>	<b>VS = 2.7V to 5.5V, VCM &lt; (V+) - 2V</b>			<b>150</b>	μV/V
Channel Separation, DC			0.4		μV/V
f = 1kHz			128		dB
<b>INPUT VOLTAGE RANGE</b>					
Common-Mode Voltage Range VCM		(V-) - 0.2		(V+) + 0.2	V
Common-Mode Rejection Ratio CMRR	(V-) - 0.2V < CVM < (V+) - 2V	80	90		dB
<b>over Temperature</b>	(V-) - 0.2V < VCM < (V+) - 2V	<b>70</b>			dB
	VS=5.5V, (V-) - 0.2V < CVM < (V+) + 0.2V	66			dB
<b>over Temperature</b>	VS=5.5V, (V-) - 0.2V < VCM < (V+) + 0.2V	60			dB
<b>INPUT BIAS CURRENT</b>					
Input Bias Current IB			±0.5	±10	pA
Input Offset Current IOS			±0.5	±10	pA
<b>INPUT IMPEDANCE</b>					
Differential			10 <sup>13</sup>   3		Ω  pF
Common-Mode			10 <sup>13</sup>   6		Ω  pF
<b>NOISE</b>	VCM < (V+) - 2V				
Input Voltage Noise, f = 0.1Hz to 10Hz			10		μVPP
Input Voltage Noise Density, f = 10kHz en			15		nV/ √Hz
Input Current Noise Density, f = 10kHz in			4		fA/√Hz
<b>OPEN-LOOP GAIN</b>					
Open-Loop Voltage Gain AOL	VS=5V, RL=100kΩ, 0.025V < VO < 4.975V	94	110		dB
<b>over Temperature</b>	<b>VS=5V, RL=100kΩ, 0.025V &lt; VO &lt; 4.975V</b>	<b>80</b>			dB
	VS=5V, RL=5kΩ, 0.025V < VO < 4.875V	94	106		dB
<b>over Temperature</b>	<b>VS=5V, RL=5kΩ, 0.025V &lt; VO &lt; 4.875V</b>	<b>80</b>			dB
<b>OUTPUT</b>					
Voltage Output Swing from Rail	RL = 100kΩ		18	25	mV
<b>over Temperature</b>	RL = 100kΩ			<b>25</b>	mV
	RL = 5kΩ		100	125	mV
<b>over Temperature</b>	RL = 5kΩ			<b>125</b>	mV
Short-Circuit Current ISC		See Typical Characteristics			
Capacitive Load Drive CLOAD		See Typical Characteristics			
Open-Loop Output Impedance	f = 1MHz, Io = 0		220		Ω

<b>FREQUENCY RESPONSE</b>		CL = 100pF				
Gain-Bandwidth Product	GBW			6.5		MHz
Slew Rate	SR	G = +1		5		V/ $\mu$ s
Settling Time, 0.1%	tS	VS = 5V, 2V Step, G = +1		1		$\mu$ s
0.01%		VS = 5V, 2V Step, G = +1		1.5		$\mu$ s
Overload Recovery Time		VIN•Gain > VS		0.3		$\mu$ s
Total Harmonic Distortion+Noise	THD+N	VS = 5V, VO = 3VPP, G = +1, f = 1kHz		0.0013		%
<b>ENABLE/SHUTDOWN</b>						
tOFF				3		$\mu$ s
tON				12		$\mu$ s
VL (shutdown)			V–		(V–)+0.8	V
VH (amplifier is active)			(V–) + 2		V+	V
Input Bias Current of Enable Pin				0.2		$\mu$ A
IQSD (per amplifier)				< 0.5	1	$\mu$ A
<b>POWER SUPPLY</b>						
Specified Voltage Range	VS		2.7		5.5	V
Operating Voltage Range				2.3to5.5		V
Quiescent Current (per amplifier)	IQ	IO = 0		585	750	$\mu$ A
over Temperature					<b>800</b>	$\mu$ A
<b>TEMPERATURE RANGE</b>						
Specified Range			–40		+125	°C
Operating Range			–55		+150	°C
Storage Range			–65		+150	°C
Thermal Resistance	$\theta$ JA					°C/W
SOT23-5, SOT23-6, SOT23-8				+200		°C/W
MSOP-10, SOP-8				+150		°C/W
SOP-14, TSSOP-14				+100		°C/W

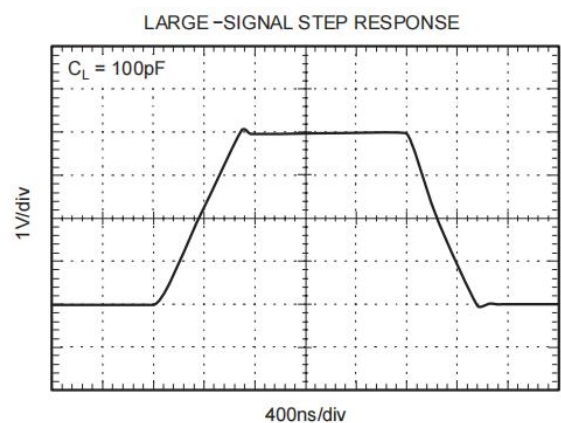
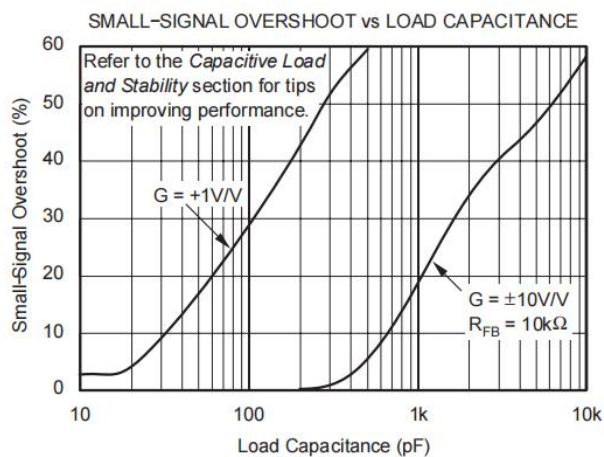
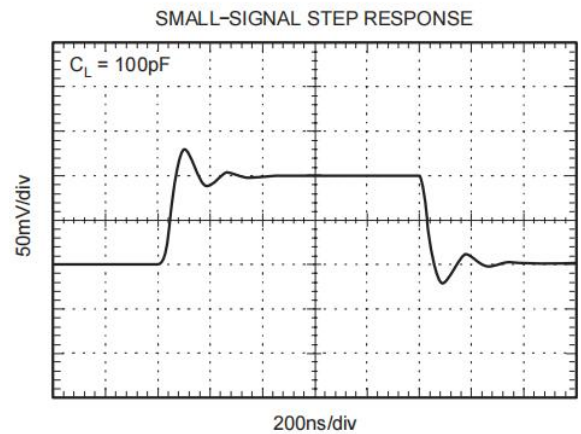
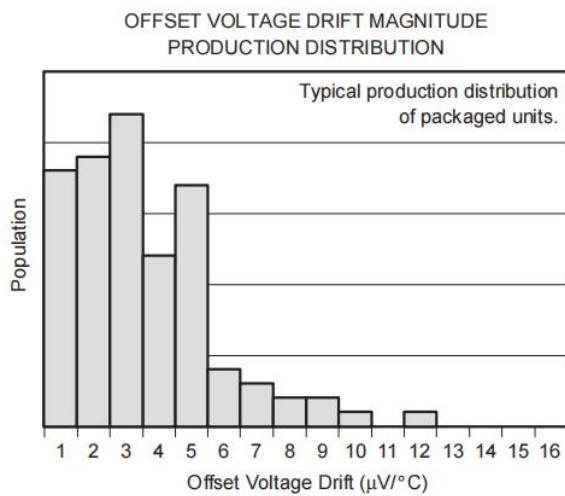
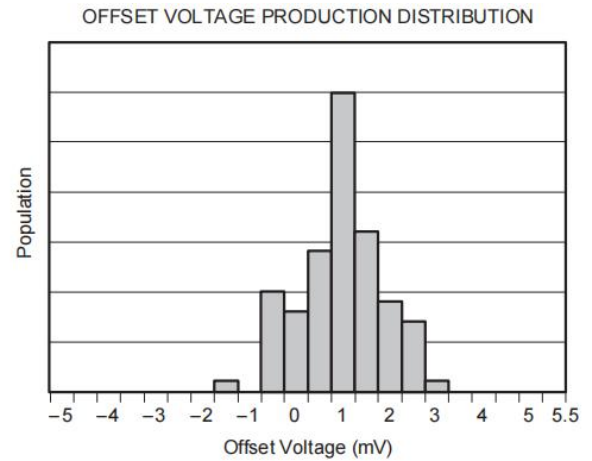
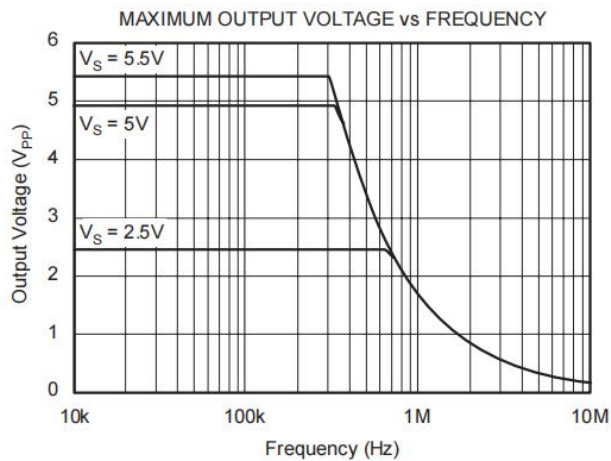
### TYPICAL CHARACTERISTICS

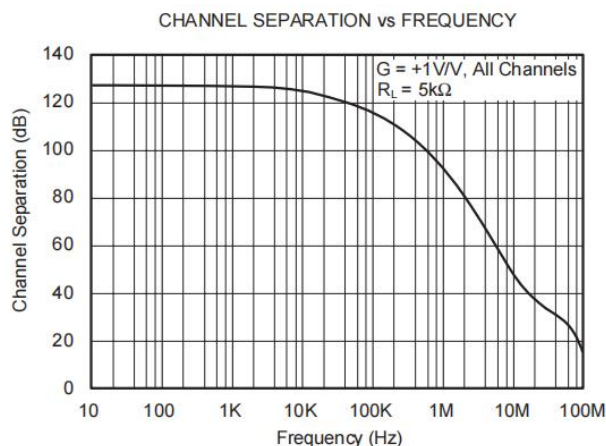
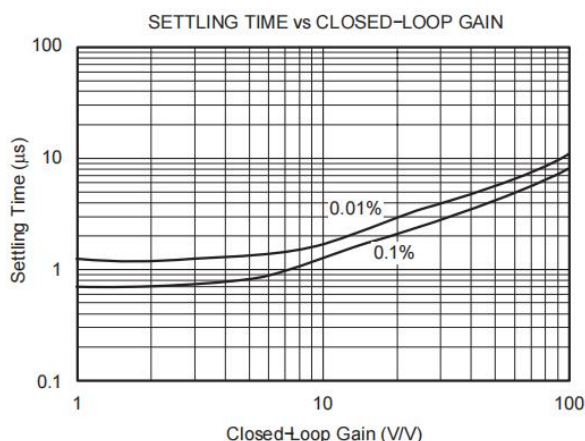
At  $T_A = +25^\circ\text{C}$ ,  $R_L = 10\text{k}\Omega$  connected to  $V_S/2$ , and  $V_{OUT} = V_S/2$ , unless otherwise noted.











## APPLICATIONS

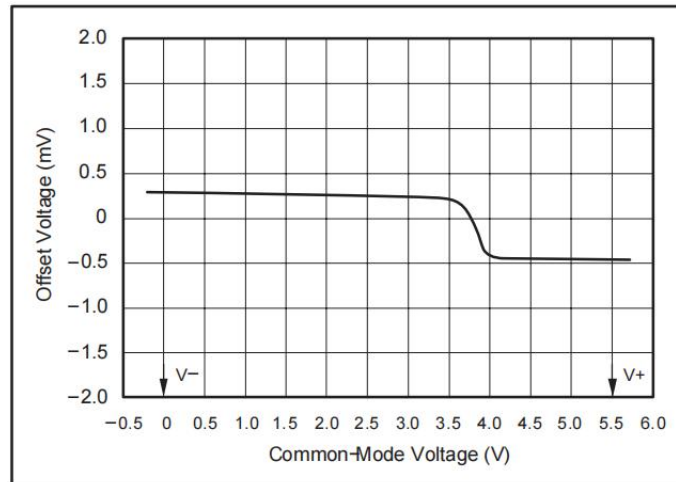
The HXA373A and HXA373 series op amps are unity-gain stable and suitable for a wide range of general-purpose applications. Rail-to-rail input and output make them ideal for driving sampling Analog-to-Digital Converters (ADCs). Excellent AC performance makes them well suited for audio applications. The class AB output stage is capable of driving 100kΩ loads connected to any point between V+ and ground. The input common-mode voltage range includes both rails, allowing the HXA373A and HXA373 series op amps to be used in virtually any single-supply application up to a supply voltage of +5.5V. Rail-to-rail input and output swing significantly increases dynamic range, especially in low-supply applications. Power-supply pins should be bypassed with 0.01μF ceramic capacitors.

## OPERATING VOLTAGE

The HXA373A and HXA373 op amps are specified and tested over a power-supply range of +2.7V to +5.5V ( $\pm 1.35V$  to  $\pm 2.75V$ ). However, the supply voltage may range from +2.3V to +5.5V ( $\pm 1.15V$  to  $\pm 2.75V$ ). Supply voltages higher than 7.0V (absolute maximum) can permanently damage the amplifier. Parameters that vary over supply voltage or temperature are shown in the Typical Characteristics section of this data sheet.

## COMMON-MODE VOLTAGE RANGE

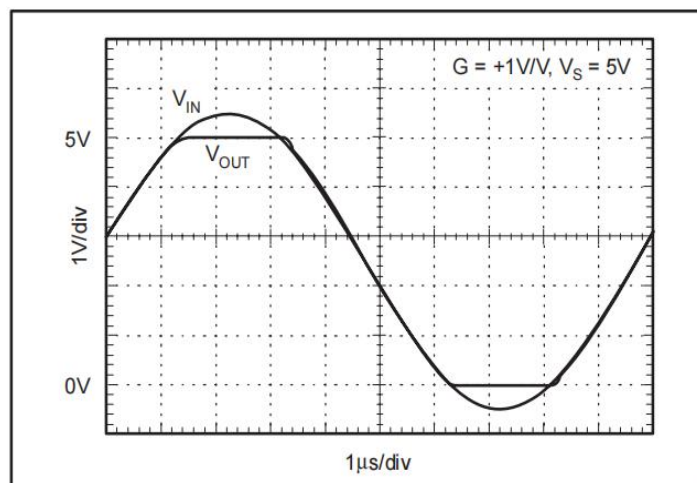
The input common-mode voltage range of the HXA373A and HXA373 series extends 200mV beyond the supply rails. This is achieved with a complementary input stage—an N-channel input differential pair in parallel with a P-channel differential pair. The N-channel pair is active for input voltages close to the positive rail, typically  $(V+) - 1.65V$  to 200mV above the positive supply, while the P-channel pair is on for inputs from 200mV below the negative supply to approximately  $(V+) - 1.65V$ . There is a 500mV transition region, typically  $(V+) - 1.9V$  to  $(V+) - 1.4V$ , in which both pairs are on. This 500mV transition region, shown in Figure 1, can vary  $\pm 300mV$  with process variation. Thus, the transition region (both stages on) can range from  $(V+) - 2.2V$  to  $(V+) - 1.7V$  on the low end, up to  $(V+) - 1.6V$  to  $(V+) - 1.1V$  on the high end. Within the 500mV transition region PSRR, CMRR, offset voltage, offset drift, and THD may be degraded compared to operation outside this region.



**Figure 1. Behavior of Typical Transition Region at Room Temperature**

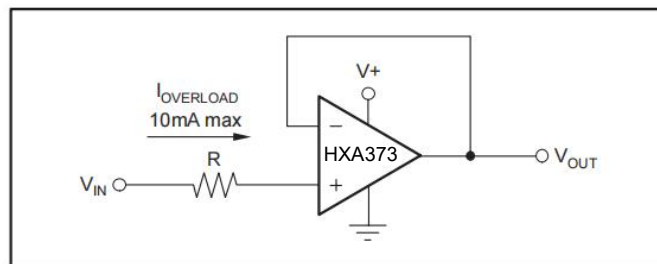
## RAIL-TO-RAIL INPUT

The input common-mode range extends from  $(V-) - 0.2V$  to  $(V+) + 0.2V$ . For normal operation, inputs should be limited to this range. The absolute maximum input voltage is 500mV beyond the supplies. Inputs greater than the input common-mode range but less than the maximum input voltage, while not valid, will not cause any damage to the op amp. Unlike some other op amps, if input current is limited, the inputs may go beyond the supplies without phase inversion, as shown in Figure 2.



**Figure 2. HXA373A: No Phase Inversion with Inputs Greater Than the Power-Supply Voltage**

Normally, input bias current is approximately 500fA; however, input voltages exceeding the power supplies by more than 500mV can cause excessive current to flow in or out of the input pins. Momentary voltages greater than 500mV beyond the power supply can be tolerated if the current on the input pins is limited to 10mA. This is easily accomplished with an input resistor; see Figure 3. (Many input signals are inherently current-limited to less than 10mA, therefore, a limiting resistor is not required.)



**Figure 3. Input Current Protection for Voltages Exceeding the Supply Voltage**

## RAIL-TO-RAIL OUTPUT

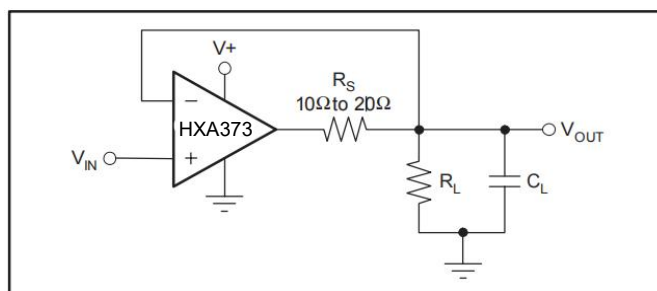
A class AB output stage with common-source transistors is used to achieve rail-to-rail output. For light resistive loads ( $>100\text{k}\Omega$ ), the output voltage can typically swing to within 18mV from the supply rails. With moderate resistive loads (5k $\Omega$  to 50k $\Omega$ ), the output can typically swing to within 100mV from the supply rails and maintain high open-loop gain. See the Typical Characteristics curve, Output Voltage Swing vs Output Current, for more information.

## CAPACITIVE LOAD AND STABILITY

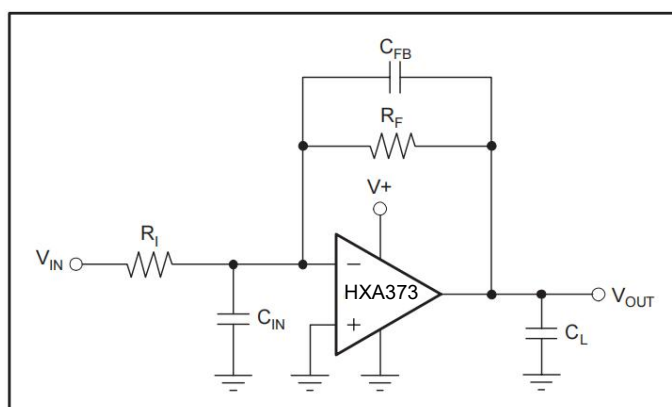
HXA373A series op amps can drive a wide range of capacitive loads. However, under certain conditions, all op amps may become unstable. Op amp configuration, gain, and load value are just a few of the factors to consider when determining stability. An op amp in unity-gain configuration is the most susceptible to the effects of capacitive load. The capacitive load reacts with the op amp output resistance, along with any additional load resistance, to create a pole in the small-signal response that degrades the phase margin. The HXA373A series op amps perform well in unity-gain configuration, with a pure capacitive load up to approximately 250pF. Increased gains allow the amplifier to drive more capacitance. See the Typical Characteristics curve, Small-Signal Overshoot vs Capacitive Load, for further details.

One method of improving capacitive load drive in the unity-gain configuration is to insert a small (10 $\Omega$  to 20 $\Omega$ ) resistor,  $R_S$ , in series with the output, as shown in Figure 4. This significantly reduces ringing while maintaining DC performance for purely capacitive loads. When there is a resistive load in parallel with the capacitive load,  $R_S$  must be placed within the feedback loop as shown to allow the feedback loop to compensate for the voltage divider created by  $R_S$  and  $R_L$ .

In unity-gain inverter configuration, phase margin can be reduced by the reaction between the capacitance at the op amp input and the gain setting resistors, thus degrading capacitive load drive. Best performance is achieved by using small valued resistors. However, when large valued resistors cannot be avoided, a small (4pF to 6pF) capacitor, CFB, can be inserted in the feedback, as shown in Figure 5. This significantly reduces overshoot by compensating the effect of capacitance,  $C_{IN}$ , which includes the amplifier input capacitance and PC board parasitic capacitance.

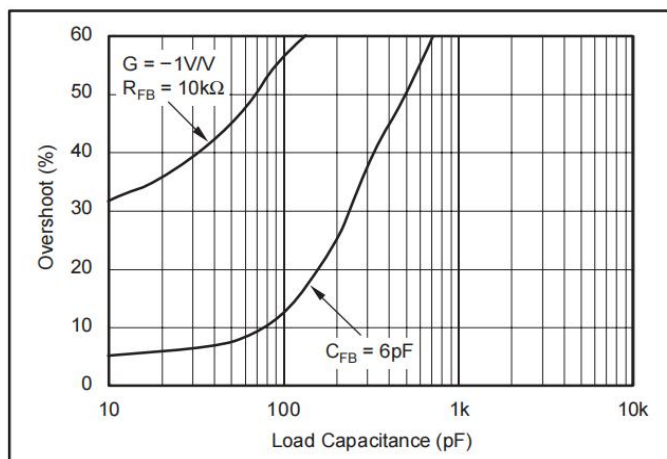


**Figure 4. Series Resistor in Unity-Gain Configuration Improves Capacitive Load Drive**



**Figure 5. Improving Capacitive Load Drive**

For example, when driving a 100pF load in unity-gain inverter configuration, adding a 6pF capacitor in parallel with the 10kΩ feedback resistor decreases overshoot from 57% to 12%, as shown in Figure 6.



**Figure 6. Improving Capacitive Load Drive**

### DRIVING ADCs

The HXA373A and HXA373 series op amps are optimized for driving medium-speed sampling ADCs. The HXA373A and HXA373 op amps buffer the ADC input capacitance and resulting charge injection, while providing signal gain.

The HXA373A is shown driving the ADS7816 in a basic noninverting configuration, as shown in Figure 7. The ADS7816 is a 12-bit, MicroPower sampling converter in the MSOP-8 package. When used with the low-power, miniature packages of the HXA373A, the combination is ideal for space-limited, low-power applications. In this configuration, an RC network at the ADC input can be used to provide anti-aliasing filtering.

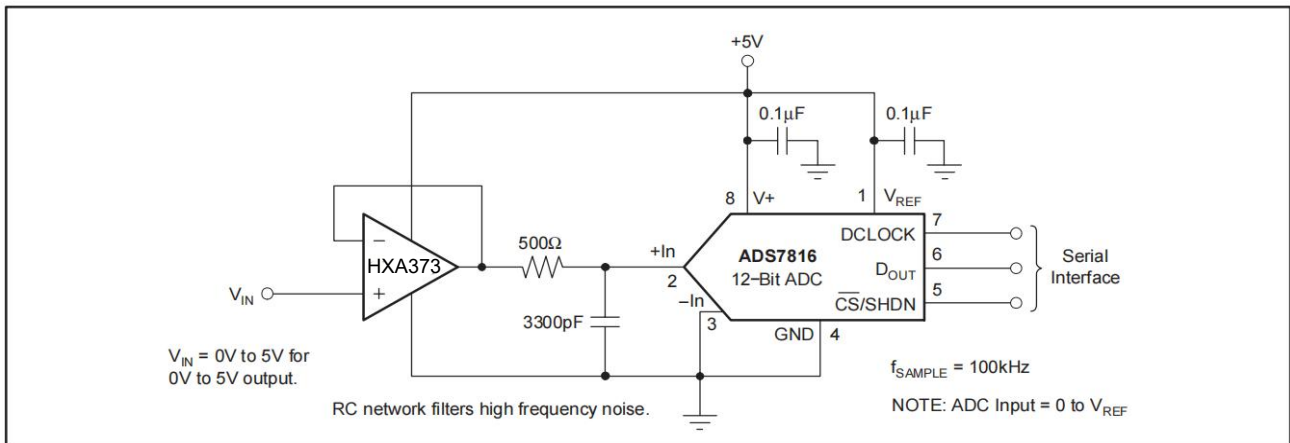
Figure 8 shows the HXA373A driving the ADS7816 in a speech band-pass filtered data acquisition system. This

small, low-cost solution provides the necessary amplification and signal conditioning to interface directly with an electret microphone. This circuit will operate with  $V_S = 2.7V$  to  $5V$ .

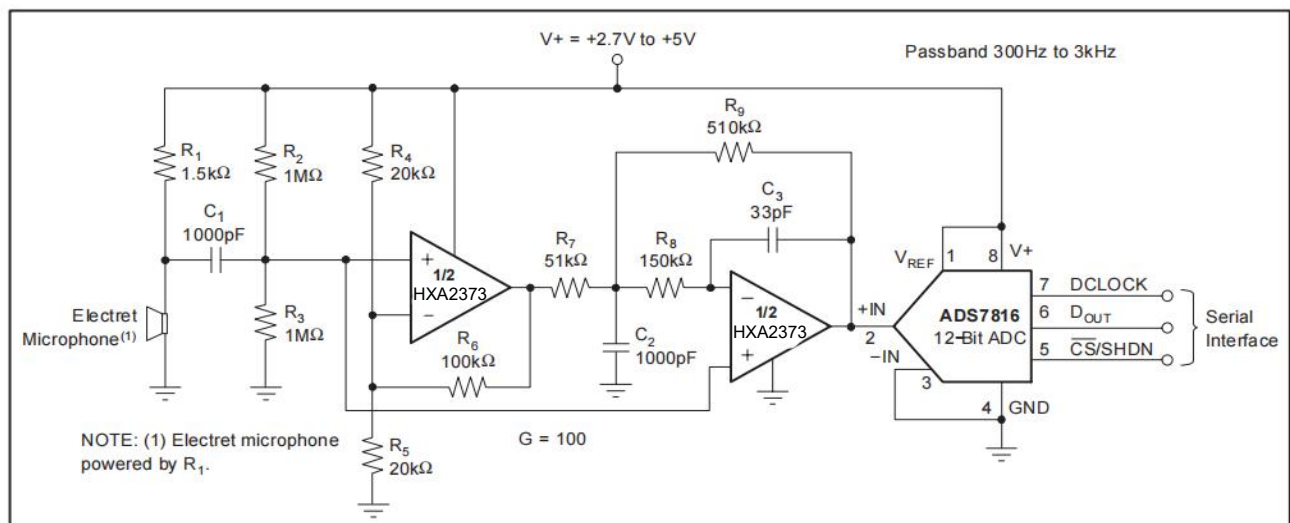
The HXA373A is shown in the inverting configuration described in Figure 9. In this configuration, filtering may be accomplished with the capacitor across the feedback resistor.

### ENABLE/SHUTDOWN

HXA373A and HXA373 series op amps typically require  $585\mu A$  quiescent current. The enable/shutdown feature of the HXA373A allows the op amp to be shut off in order to reduce this current to less than  $1\mu A$ .

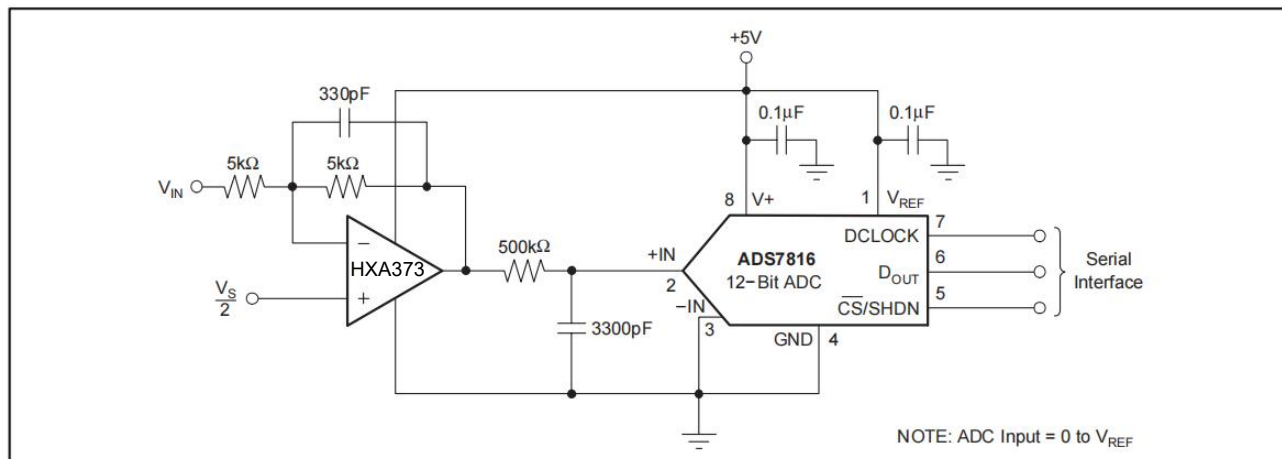


**Figure 7. The HXA373A in Noninverting Configuration Driving the ADS7816**

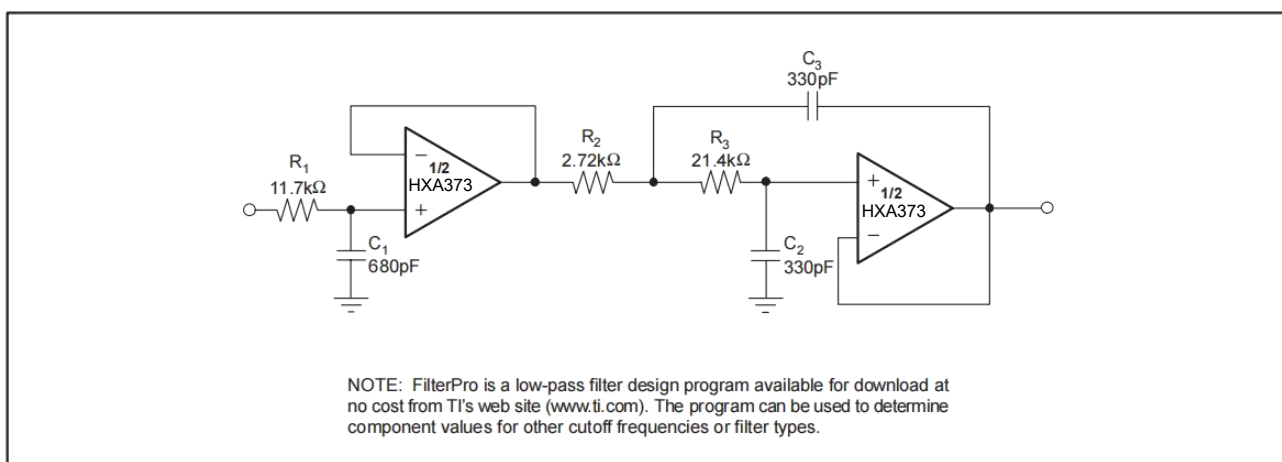


**Figure 8. The HXA2373 as a Speech Bypass Filtered Data Acquisition System**





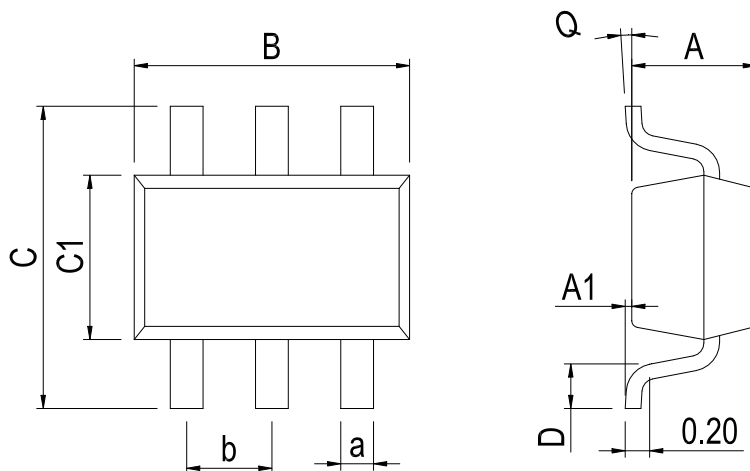
**Figure 9. The HXA373A in Inverting Configuration Driving the ADS7816**



**Figure 10. Three-Pole Sallen-Key Butterworth Low-Pass Filter**

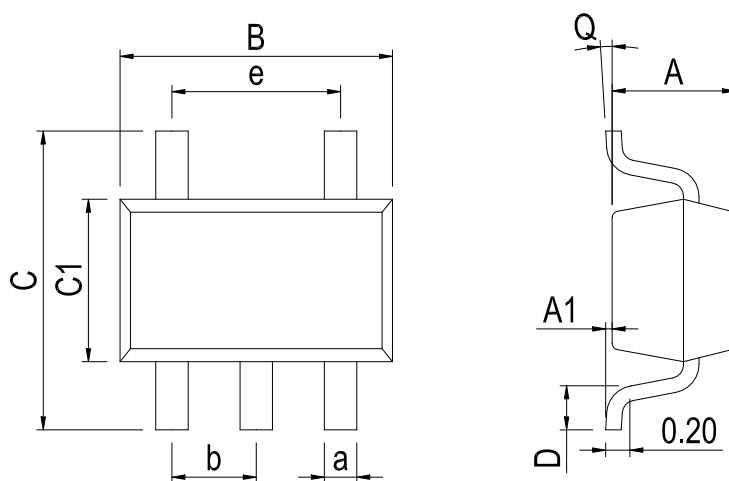
## PHYSICAL DIMENSIONS

### SOT-23-6



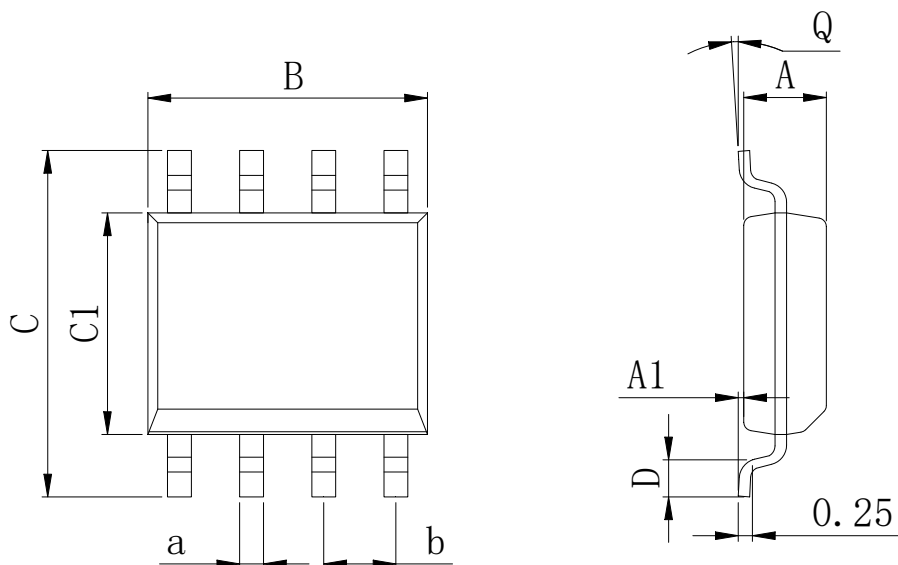
Dimensions In Millimeters(SOT23-6)									
Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	1.00	0.00	2.82	2.65	1.50	0.30	0°	0.30	0.95 BSC
Max:	1.15	0.15	3.02	2.95	1.70	0.60	8°	0.50	

### SOT-23-5

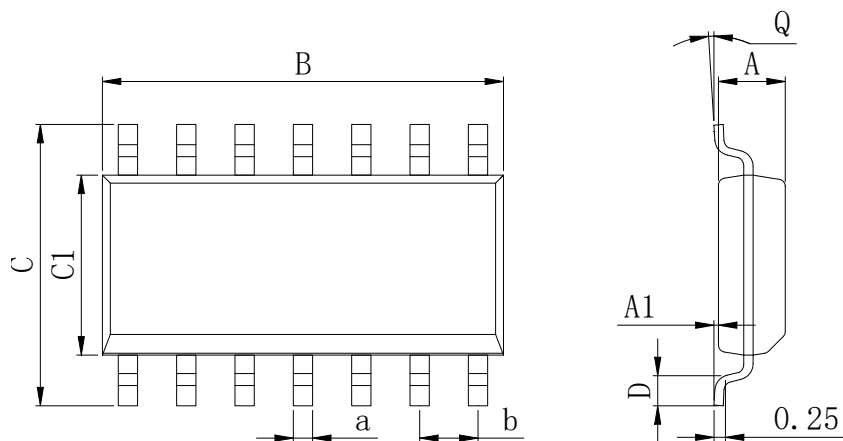


Dimensions In Millimeters(SOT-23-5)										
Symbol:	A	A1	B	C	C1	D	Q	a	b	e
Min:	1.00	0.00	2.82	2.65	1.50	0.30	0°	0.30	0.95 BSC	1.90 BSC
Max:	1.15	0.15	3.02	2.95	1.70	0.60	8°	0.50		

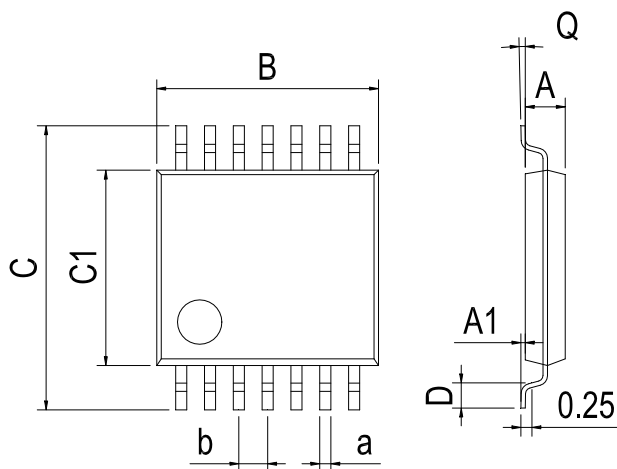


**SOP-8**


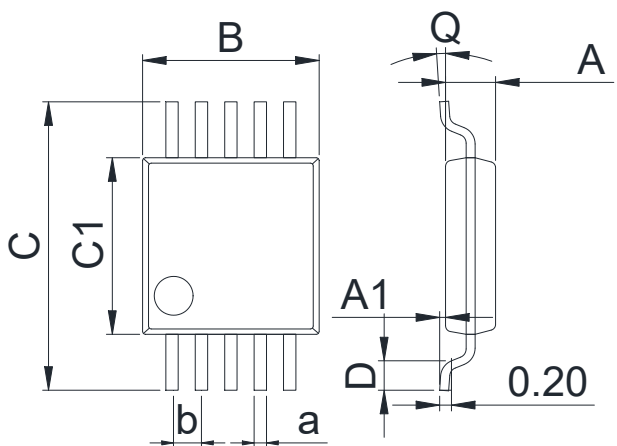
Dimensions In Millimeters(SOP-8)									
Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	1.35	0.05	4.90	5.80	3.80	0.40	0°	0.35	1.27 BSC
Max:	1.55	0.20	5.10	6.20	4.00	0.80	8°	0.45	

**SOP-14**


Dimensions In Millimeters(SOP-14)									
Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	1.35	0.05	8.55	5.80	3.80	0.40	0°	0.35	1.27 BSC
Max:	1.55	0.20	8.75	6.20	4.00	0.80	8°	0.45	

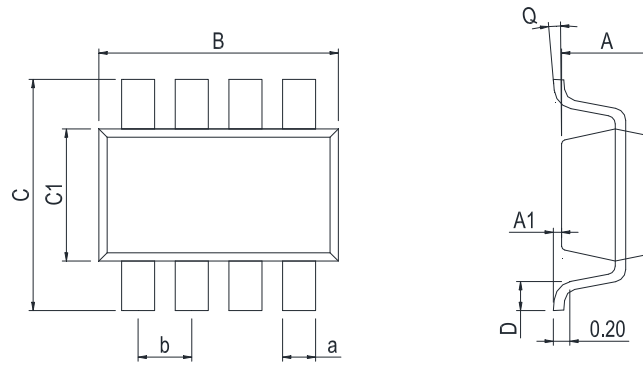
**TSSOP-14**


Dimensions In Millimeters(TSSOP-14)									
Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	0.85	0.05	4.90	6.20	4.30	0.40	0°	0.20	0.65 BSC
Max:	0.95	0.20	5.10	6.60	4.50	0.80	8°	0.25	

**MSOP-10**


Dimensions In Millimeters(MSOP-10)									
Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	0.80	0.05	2.90	4.75	2.90	0.35	0°	0.17	0.5 BSC
Max:	0.90	0.20	3.10	5.05	3.10	0.75	8°	0.33	

## SOT-23-8



Dimensions In Millimeters(SOT-23-8)									
Symbol:	A	A1	B	C	C1	D	Q	a	b
Min:	1.05	0.00	2.82	2.65	1.50	0.30	0°	0.30	0.65 BSC
Max:	1.15	0.15	3.02	2.95	1.70	0.60	8°	0.40	

**REVISION HISTORY**

REVISION NUMBER	DATE	REVISION	PAGE
V1.0	2018-8	New	1-21
V1.1	2023-7	Update encapsulation type、Update Lead Temperature	1、 4
V1.2	2024-11	Update SOT-23-5 and SOT-23-6 Physical dimension	16

**IMPORTANT STATEMENT:**

Hanschip Semiconductor reserves the right to change its products and services without notice. Before ordering, the customer shall obtain the latest relevant information and verify whether the information is up to date and complete. Hanschip Semiconductor does not assume any responsibility or obligation for the altered documents.

Customers are responsible for complying with safety standards and taking safety measures when using Hanschip Semiconductor products for system design and machine manufacturing. You will bear all the following responsibilities: select the appropriate Hanschip Semiconductor products for your application; Design, validate and test your application; Ensure that your application meets the appropriate standards and any other safety, security or other requirements. To avoid the occurrence of potential risks that may lead to personal injury or property loss.

Hanschip Semiconductor products have not been approved for applications in life support, military, aerospace and other fields, and Hanschip Semiconductor will not bear the consequences caused by the application of products in these fields. All problems, responsibilities and losses arising from the user's use beyond the applicable area of the product shall be borne by the user and have nothing to do with Hanschip Semiconductor, and the user shall not claim any compensation liability against Hanschip Semiconductor by the terms of this Agreement.

The technical and reliability data (including data sheets), design resources (including reference designs), application or other design suggestions, network tools, safety information and other resources provided for the performance of semiconductor products produced by Hanschip Semiconductor are not guaranteed to be free from defects and no warranty, express or implied, is made. The use of testing and other quality control technologies is limited to the quality assurance scope of Hanschip Semiconductor. Not all parameters of each device need to be tested.

The documentation of Hanschip Semiconductor authorizes you to use these resources only for developing the application of the product described in this document. You have no right to use any other Hanschip Semiconductor intellectual property rights or any third party intellectual property rights. It is strictly forbidden to make other copies or displays of these resources. You should fully compensate Hanschip Semiconductor and its agents for any claims, damages, costs, losses and debts caused by the use of these resources. Hanschip Semiconductor accepts no liability for any loss or damage caused by infringement.